

# A High-gain and Low-scattering Waveguide Slot Antenna of Artificial Magnetic Conductor Octagonal Ring Arrangement

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**Abstract.** A novel design of high-gain and low-scattering waveguide slot antenna is proposed in this paper. Firstly the scattering pattern of artificial magnetic conductor (AMC) composite surface is estimated by array factor analysis method. The comparison between octagonal ring arrangement and chessboard arrangement proves that the former arrangement has the characteristic of diffuseness-like and expands the bandwidth of radar cross section (RCS) reduction. Secondly, the metal surface of waveguide slot antenna (WSA) is replaced by the octagonal ring arrangement composite surface (ORACS). The gain is improved because of spurious radiation units which are around the slot. At the same time using the phase cancellation principle, a backscatter null achieves RCS reduction in the vertical direction. Experimental results show that for the novel antenna after loading with the ORACS, the gain is improved by 5 dB; the bandwidth of RCS reduction (reduction greater than 10 dB) is 5.24–5.92 GHz.

## Keywords

Waveguide slot antenna, artificial magnetic conductor, octagonal ring arrangement

## 1. Introduction

With rapid development of stealth technology, RCS has become a major specification of aircrafts. Antenna which is an indispensable part of the wireless communication system is also the main contributor of the total RCS of the aircraft. It has practical significance to reduce the RCS of the aircraft and improve the radiation performance of the antennas.

In 1999, D. Sievenpiper proposed a new 2-D metamaterial, Artificial Magnetic Conductor [1], [2]. When plane wave arrives on the structure, the phase of reflect wave keeps unchanged within a particular frequency band. In [3], [4], AMC is used as the back-cavity or cladding to achieve the low-profile antennas. In order to improve the gain of antenna, AMC is loaded on the aperture, [5–8]. In

[5], the AMC is loaded around the rectangular microstrip antenna, the gain is improved in consequence of the change of surface current. In [9], an ultra-thin absorber is manufactured by loading with lump resistance. In [10], the metal surface of WSA is replaced by the ultra-thin absorber, and the gain is decreased in consequence of loading with the lump resistance.

In 2007, Paquay proposed a novel structure which is combined of AMC and PEC by chessboard arrangement [12]. Using the phase cancellation principle, the RCS is reduced in specified space. Even without the loss component, through arranging properly, the purpose of gain enhancement and RCS reduction is achieved. In [13], the composite surface is used where AMC and PEC are arranged by square ring and the WSA are integrated. The gain is improved by 1.7 dB, and the fractional bandwidth of RCS reduction (reduction greater than 10 dB) is 8.5%. In [14], the composite surface which is combined of AMC and PEC by chessboard arrangement is loaded on the WSA, the gain is improved by 5.1 dB, and the fractional bandwidth is 5.8%. In [15], the fractional bandwidth of RCS reduction gets extended by changing the cell, but the gain is improved limitedly by 3 dB.

In this paper, a high-gain low-scattering waveguide slot antenna of AMC octagonal ring arrangement is optimized and simulated. Experiments are carried out to verify the simulation results, and the experimental results show that the gain of the antenna is improved by 5 dB, and the fractional bandwidth of RCS reduction is 12.3%.

## 2. Analysis of ORACS

ORACS is a composite surface which is combined PEC and AMC by octagonal ring arrangement. Compared with chessboard arrangement, more backscatter beams are achieved due to the irregular arrangement.

As the plane wave arrives on the surface, the amplitude of the surface current is equal while the phase is inverse; the scattering field is described as:

$$E_{PEC} = A \cdot \exp(j\theta_1) \cdot S_{PEC} \quad (1)$$

$$E_{AMC} = A \cdot \exp(j\theta_2) \cdot S_{AMC} \quad (2)$$

The total scattering field is described as:

$$E = E_{PEC} \cdot A1 + E_{AMC} \cdot A2 \quad (3)$$

in which  $S$  represents areas of the AMC and PEC,  $\theta_1 \approx 180^\circ$ ,  $\theta_2 \approx 0^\circ$ ,  $A1$  and  $A2$  are the parameters of the array.

An AMC cell is designed as ‘0’ element with a 0deg phase response, and PEC is designed as ‘1’ element with a 180deg phase response. The composite surfaces are shown in Fig. 1. Elements are regarded as single cells in

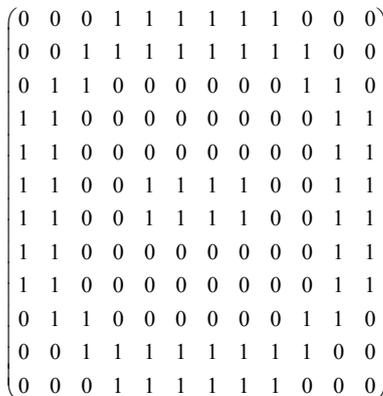
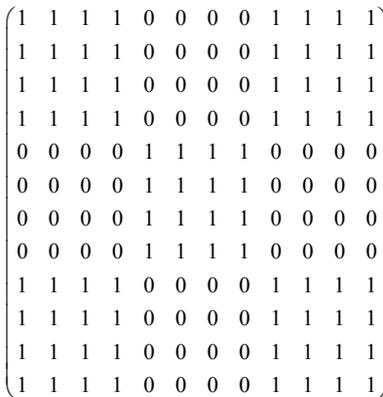
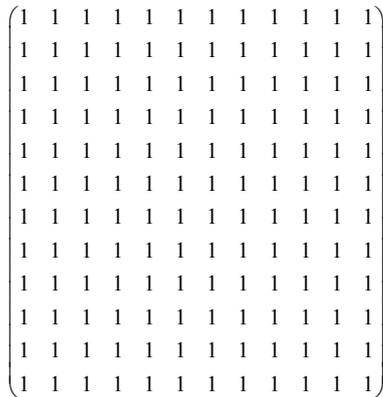


Fig. 1. Diagram of composite structure reflection phase. (a) PEC, (b) Chessboard arrangement, (c) Octagonal ring arrangement.

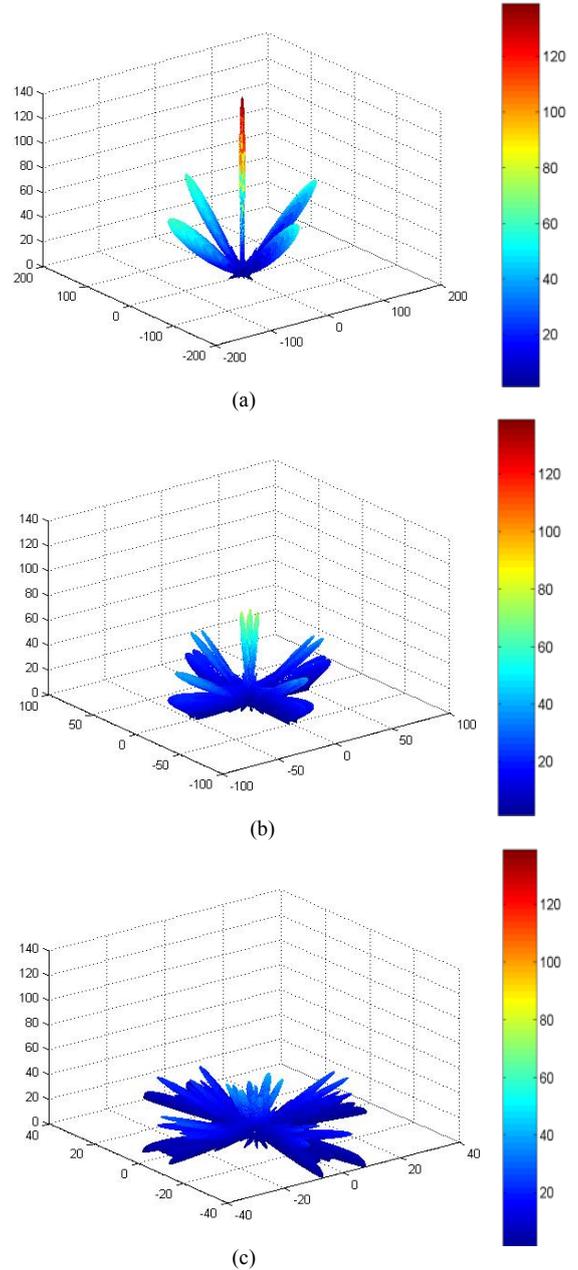


Fig. 2. Three-dimensional scattering array factor. (a) PEC, (b) Chessboard arrangement, (c) Octagonal ring arrangement.

an antenna array, the amplitude is 1, and the distance between the cells is  $0.15\lambda$ . The reflection phase information and cell location information are translated into antenna array information. The scattering patterns of three composite surfaces are analyzed by using the method of array factor [16], shown as Fig. 2.

Shown as Fig. 2, the reflect energy of octagonal ring arrangement is scattered into more space than the chessboard arrangement, and the characteristic of diffuseness-like is proved.

In order to illustrate the characteristic of extending the band of RCS reduction, the method of array factor analysis is taken again. The amplitude of RCS reduction is observed by picking up the value of array factor in  $+z$  axis. When the

'0' element is designed as an AMC cell with a 30deg phase response, '1' keeps unchanged in Fig. 1, 144, 8, 1.68 represent the value of PEC, chessboard arrangement and octagonal ring arrangement respectively. When the '0' element is designed as an AMC cell with a -30deg phase response, '1' keeps unchanged, 144, 16, 12.38 represent the value of PEC, chessboard arrangement and octagonal ring arrangement respectively. The value of array factor in the +z axis of different arrangement is shown in Fig. 3. By comparison the value of three arrangements, it is obvious that the octagonal ring arrangement is significantly less than chessboard arrangement and PEC in the same frequency band, that is, the bandwidth of RCS reduction is expanded.

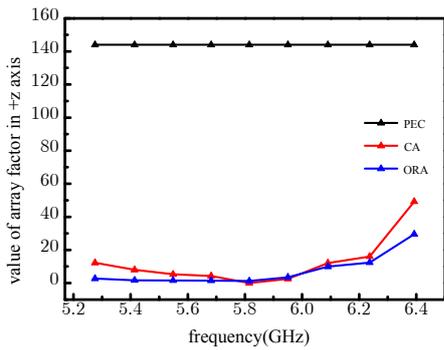


Fig. 3. Value of array factor in +z axis of different arrangement.

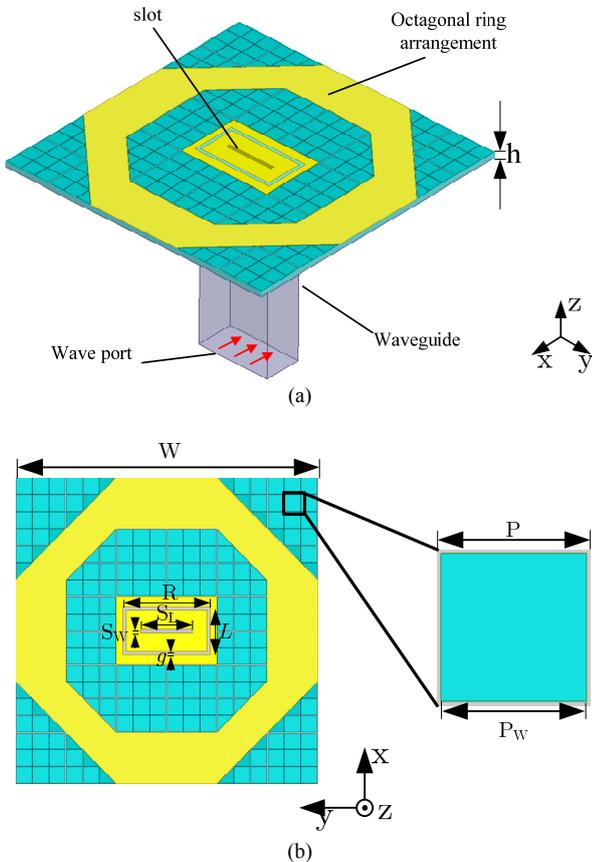


Fig. 4. WSA loading with ORACS. (a) Overall, (b) ORACS and single unit.

### 3. Simulation of the WSA Loading with ORACS

Figure 4 gives the scheme of WSA loading with ORACS. The metal surface is replaced by ORACS, which become a part of aperture, and the metal ground keeps un-changed. The sizes after optimizing are:  $W = 150$  mm,  $h = 3$  mm,  $R = 42$  mm,  $L = 22$  mm,  $g = 1$  mm,  $S_L = 26$  mm,  $S_W = 2$  mm,  $P = 8.2$  mm,  $P_w = 7.9$  mm.

#### 3.1 Analysis of Scattering Characteristic of ORACS

The different loading methods and the RCS reduction curves of WSA are reported in Fig. 5 and Fig. 6. Shown as Fig. 6, the results prove that both the arrangements work well to reduce RCS. The RCS reduction bandwidth of loading B is narrower than loading C, and RCS reduction value of loading C is significantly less than loading B.

Figure 7 shows the scattering pattern of loading C, the energy is diffused into much space, which agrees with the former prediction of the array factor method.

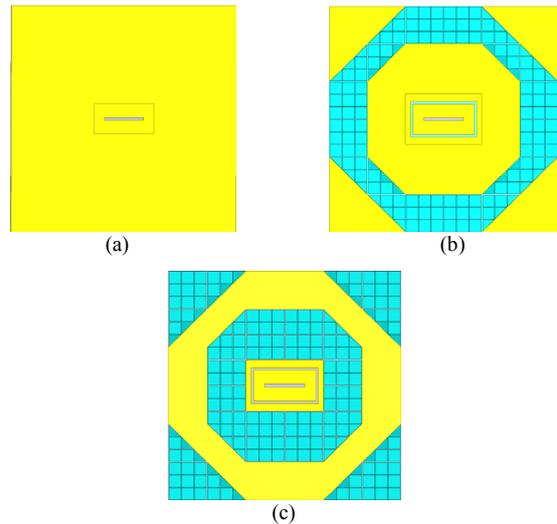


Fig. 5. ORACS of three loading methods. (a) Loading A, (b) Loading B, (c) Loading C.

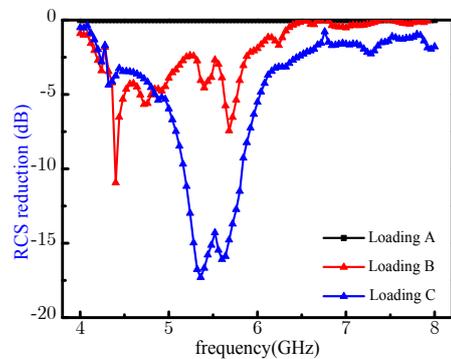


Fig. 6. Simulation of RCS reduction curve under normal incident wave.

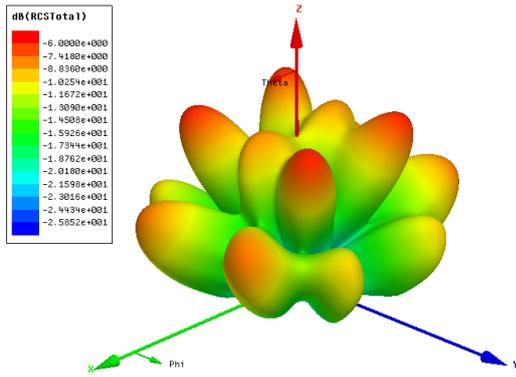


Fig. 7. Loading C's three-dimensional scattering pattern under normal incident wave.

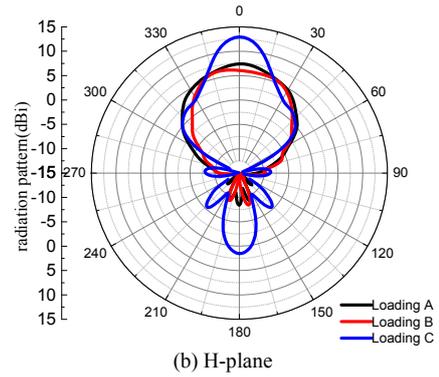


Fig. 9. The pattern of the three loading methods at the resonant frequency.

### 3.2 Analysis of Radiation Performance of WSA Loading with ORACS

Figure 8 and 9 show the  $S_{11}$  curves and the radiation patterns of WSA loading with different ORACS. The performance of  $S_{11}$  keeps unchanged after loading with the ORACS. The beam width of loading B is wider than the original antenna, and the gain is decreased. The gain of loading C is improved, and the beam width is narrowed. The maximum gain and the resonant frequency are given in Tab. 1.

Loading	A	B	C
Resonant frequency (GHz)	5.66	5.66	5.66
Gain(dB)	7.38	6.99	13.11

Tab. 1. The maximum gain and the resonant frequency.

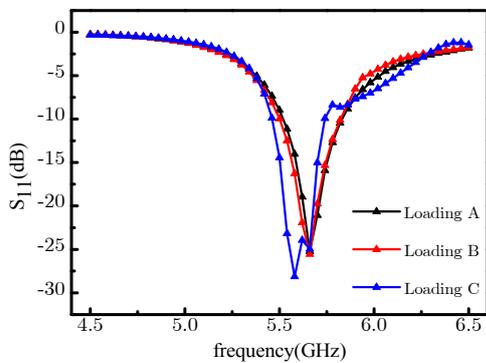


Fig. 8. The reflection coefficient of the three loading methods.

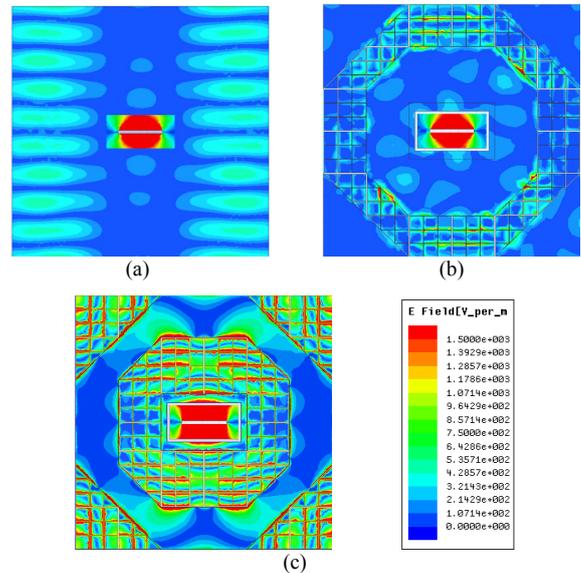


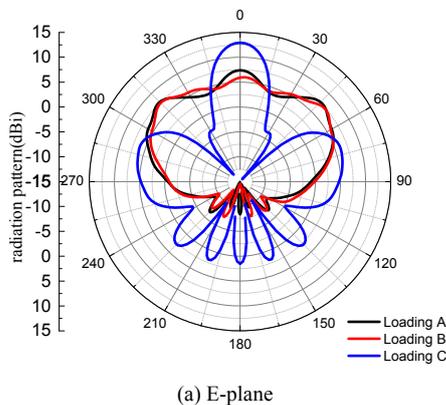
Fig. 10. Surface E-field distribution of the three loading methods at 5.66 GHz in same scale (0-1500 V-per-meter). (a) loading A, (b) loading B, (c) loading C.

Figure 10 shows the surface current distribution of three loading methods. For loading B, the gain is decreased because the AMC cells are far from the slot, and spurious radiation units are not excited effectively. For loading C, the AMC cells are around the slot, and the spurious radiation units which are excited strongly form an array with the slot, and the gain gets improved greatly.

### 4. Fabrication and Measurement

To verify the simulation results, the WSA of loading C is illustrated in Fig. 11 and has been experimentally studied. The proposed ORACS was fabricated using an optical lithographic process on a 3-mm-thick F4B-2 substrate with  $\epsilon_r = 2.65$  and fixed on the WSA by four screws.

The  $S_{11}$  curve was measured using Agilent N5230C vector network analyzer, shown in Fig. 12. Comparing to the curve of the reference antenna (Loading A), the curve of the improved antenna (Loading C) is almost the same. The resonant frequency of the reference antenna and the



improved antenna are both at 5.66 GHz and the bandwidths are 5.48–5.84 GHz and 5.5 GHz–5.84 GHz, respectively.

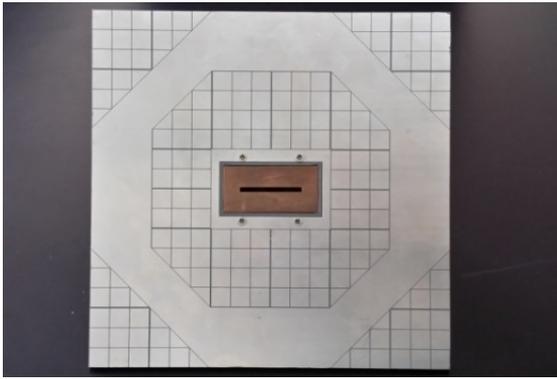


Fig. 11. Antenna sample.

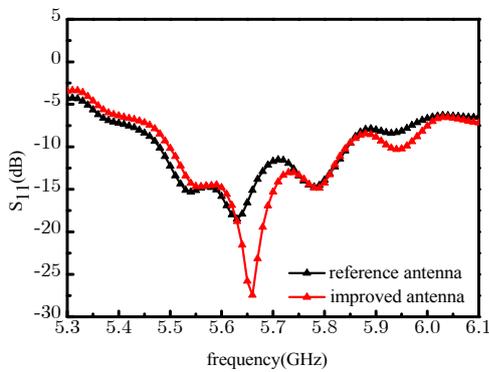
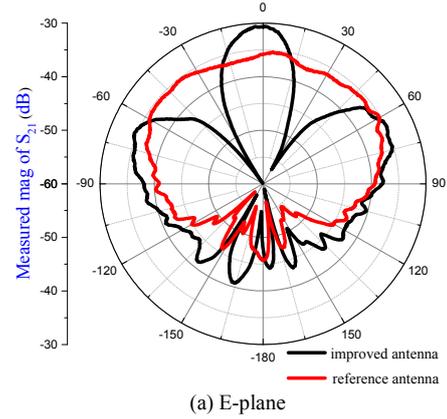
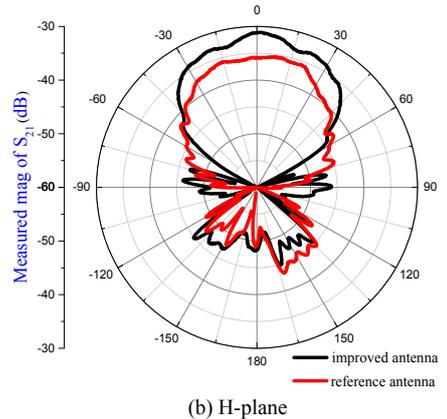


Fig. 12. The measured reflection coefficient curve.



(a) E-plane

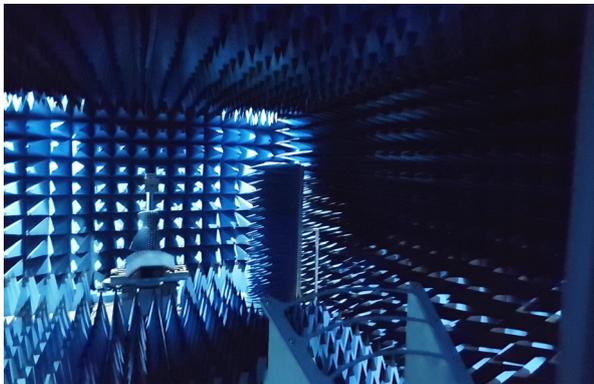


(b) H-plane

Fig. 14. The measured radiation pattern at the resonant frequency.



(a) Monostatic RCS.



(b) Radiation pattern.

Fig. 13. Measurement setup.

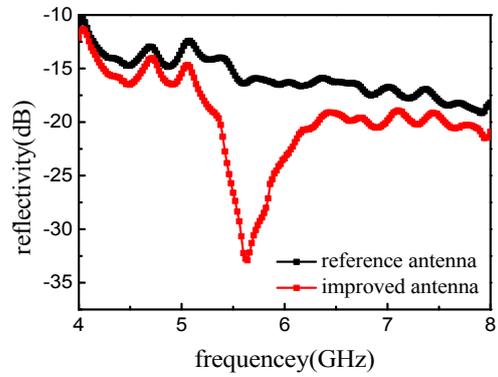


Fig. 15. Measured RCS reduction curve.

The experimental environment is shown in Fig. 13. The radiation pattern is measured in a microwave anechoic chamber and depicted in Fig. 14. The distance between the measured antenna and the horn antenna is 2.3 m ( $\geq 10\lambda$ ), satisfying the far field requirement. The measured results show that gain is enhanced by 5 dB in both E-plane and H-plane for the reference antenna.

The result of measured RCS reduction curve is shown in Fig. 15. The RCS reduction peak is 16.62 dB at 5.64 GHz. 10 dB reflection reduction is achieved from 5.24 GHz to 5.92 GHz which correspond to a fractional value of 12.3%. Considering fabrication and measurement tolerance, the measured results show accordance with the simulated ones.

These measurements above validate that our proposed design shows excellent performance in the two aspects of radiation and scattering.

In Tab. 2, the radiation and scattering performance of four kinds of WSA loading with composite surface are summarized.

	[13]	[14]	[15]	This paper
Resonant frequency (GHz)	7.0	5.7	5.66	5.66
Size of the aperture (mm)	70 (1.5 $\lambda$ )	126 (2.39 $\lambda$ )	120 (2.26 $\lambda$ )	150 (2.83 $\lambda$ )
Gain enhancement (dBi)	1.7	5.1	3.2	5
The fractional bandwidth of RCS reduction ( $\geq 10$ dB), (%)	8.5	5.8	20.0	12.3

**Tab. 2.** Comparison of design performance of WSA loading with different AMC composite surfaces.

## 5. Conclusion

In this paper, ORACS is loaded on the WSA to enhance gain and achieve the low-scattering characteristic. Simulated S-parameters, radiation patterns, gain, directivity, and RCS reduction of the WSA are compared with the measured results. Measured results indicate that the RCS of the WSA has been dramatically reduced and the gain enhancement is 5 dB. Experimentally and theoretically, it can be observed that the proposed WSA has the advantages of the gain-enhancement and RCS reduction.

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